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TLC591x 8-Channel Constant-Current LED Sink Drivers

Technical [Documents](#page-28-0)

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-
- 3.3-V or 5-V Supply Voltage
- • Maximum LED Voltage 20-V
- Thermal Shutdown for Overtemperature

2 Applications

- General LED Lighting Applications
- LED Display Systems
- LED Signage
- Automotive LED Lighting
- **White Goods**
- Gaming Machines/Entertainment

1 Features 3 Description

Tools & **[Software](#page-28-0)**

¹ The TLC591x Constant-Current LED Sink Drivers are • Eight Constant-Current Output Channels designed to work alone or cascaded. Since each Output Current Adjusted Through Single External output is independently controlled, they can be
Resistor output is independently controlled, they can be been the best for the bight Resistor
Constant Output Current Range: 3-mA to 120-mA LED voltage (VLED) allows for the use of a single LED voltage (VLED) allows for the use of a single per Channel LED per output or multiple LEDs on a single string. • Constant Output Current Invariant to Load Voltage With independently controlled outputs supplied with constant current, the LEDs can be combined in constant current, the LEDs can be combined in

parallel to create higher currents on a single string.

Open Load, Short Load and Overtemperature The constant sink current for all channels is set • Open Load, Short Load and Overtemperature The constant sink current for all channels is set through a single external resistor. This allows ²⁵⁶-Step Programmable Global Current Gain different LED drivers in the same application to sink
Fixellest Qutput Current Accureau Excellent Output Current Accuracy:
implementation of multi-color LEDs. An additional
advantage of the independent outputs is the ability to advantage of the independent outputs is the ability to $-$ Between ICs: $<$ \pm 6% (Maximum) leave unused channels floating. The flexibility of the Fast Response of Output Current **Example 2018** TLC591x LED drivers is ideal for applications such as (but not limited to): 7-segment displays, scrolling Frequency and the militar color of the superversion of the second state of the second single color displays, gaming machines, white goods,
Schmitt-Trigger Input single color displays, gaming machines, white goods, video billboards and video panels.

Support & **[Community](#page-28-0)**

으리

(1) For all available packages, see the orderable addendum at

Single Implementation of TLC5916 / TLC5917 Device

Table of Contents

4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Product Folder Links: *[TLC5916](http://www.ti.com/product/tlc5916?qgpn=tlc5916) [TLC5917](http://www.ti.com/product/tlc5917?qgpn=tlc5917)*

5 Device Comparison Table

(1) The device has one single error register for all these conditions (one error bit per channel).

6 Pin Configuration and Functions

16-PIN

Pin Functions

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7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

7.2 ESD Ratings

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

7.3 Recommended Operating Conditions

7.4 Thermal Information

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](http://www.ti.com/lit/pdf/spra953).

7.5 Electrical Characteristics: $V_{DD} = 3 V$

 $V_{DD} = 3 V$, $T_J = -40$ °C to 125°C (unless otherwise noted)

(1) Typical values represent the likely parametric nominal values determined at the time of characterization. Typical values depend on the application and configuration and may vary over time. Typical values are not ensured on production material.

(2) Specified by design.

7.6 Electrical Characteristics: $V_{DD} = 5.5 V$

 V_{DD} = 5.5 V, T_J = -40°C to 125°C (unless otherwise noted)

(1) Typical values represent the likely parametric nominal values determined at the time of characterization. Typical values depend on the application and configuration and may vary over time. Typical values are not ensured on production material.

(2) Specified by design.

7.7 Switching Characteristics: $V_{DD} = 3 V$

 $V_{DD} = 3 V$, $T_J = -40^{\circ}$ C to 125°C (unless otherwise noted)

(1) Typical values represent the likely parametric nominal values determined at the time of characterization. Typical values depend on the application and configuration and may vary over time. Typical values are not ensured on production material.

(2) If the devices are connected in cascade and t_r or t_f is large, it may be critical to achieve the timing required for data transfer between two cascaded devices.

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7.8 Switching Characteristics: $V_{DD} = 5.5 V$

 V_{DD} = 5.5 V, T_J = -40°C to 125°C (unless otherwise noted)

(1) Typical values represent the likely parametric nominal values determined at the time of characterization. Typical values depend on the application and configuration and may vary over time. Typical values are not ensured on production material.

(2) If the devices are connected in cascade and t_r or t_f is large, it may be critical to achieve the timing required for data transfer between two cascaded devices.

7.9 Timing Requirements

 $V_{DD} = 3$ V to 5.5 V (unless otherwise noted)

7.10 Typical Characteristics

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8 Parameter Measurement Information

Figure 6. Test Circuit for Switching Characteristics

Parameter Measurement Information (continued)

Figure 7. Normal Mode Timing Waveforms

ISTRUMENTS

EXAS

Figure 9. Reading Error Status Code Timing Waveforms

9 Detailed Description

9.1 Overview

The TLC591x is designed for LED displays and LED lighting applications with constant-current control and openload, shorted-load, and overtemperature detection. The TLC591x contains an 8-bit shift register and data latches, which convert serial input data into parallel output format. At the output stage, eight regulated current ports are designed to provide uniform and constant current for driving LEDs within a wide range of LED Forward Voltage (VF) variations. Used in system design for LED display applications, for example, LED panels, it provides great flexibility and device performance. Users can adjust the output current from 3 mA to 120 mA per channel through an external resistor, R_{ext} , which gives flexibility in controlling the light intensity of LEDs. The devices are designed for up to 20 V at the output port. The high clock frequency, 30 MHz, also satisfies the system requirements of high-volume data transmission.

The TLC591x provides two operation modes: Normal Mode and Special Mode. Normal mode is used for shifting LED data into and out of the driver. Special Mode includes two functions: Error Detection and Current Gain Control. The two operation modes include three phases: Normal Mode phase, Mode Switching transition phase, and Special Mode phase. The signal on the multiple function pin OE(ED2) is monitored to determine the mode. When a one-clock-wide pulse appears on $\overline{OE}(\text{ED2})$, the device enters the Mode Switching phase. At this time, the voltage level on LE(ED1) determines which mode the TLC591x switches to.

In the Normal Mode phase, the serial data can be transferred into TLC591x through the pin SDI, shifted in the shift register, and transferred out via the pin SDO. LE(ED1) can latch the serial data in the shift register to the output latch. OE(ED2) enables the output drivers to sink current.

In the Special Mode phase, the low-voltage-level signal on $\overline{OE(ED2)}$ can enable output channels and detect the status of the output current to determine if the driving current level is sufficient. The detected Error Status is loaded into the 8-bit shift register and shifted out via the pin SDO, synchronous to the CLK signal. The system controller can read the error status and determine if the LEDs are properly lit.

In the Special Mode phase, the TLC591x allows users to adjust the output current level by setting a runtimeprogrammable Configuration Code. The code is sent into the TLC591x through SDI. The positive pulse of LE(ED1) latches the code in the shift register into a built-in 8-bit configuration latch, instead of the output latch. The code affects the voltage at the terminal R-EXT and controls the output-current regulator. The output current can be finely adjusted by a gain ranging from 1/12 to 127/128 in 256 steps. Therefore, the current skew between ICs can be compensated within less than 1%. This feature is suitable for white balancing in LED color display panels.

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9.2 Functional Block Diagram

9.3 Feature Description

9.3.1 Open-Circuit Detection Principle

The LED Open-Circuit Detection compares the effective current level I_{out} with the open load detection threshold current $I_{\text{OUT,Th}}$. If I_{OUT} is below the $I_{\text{OUT,Th}}$ threshold, the TLC591x detects an open-load condition. This error status can be read as an error status code in the Special Mode. For open-circuit error detection, a channel must be on.

(1) $I_{\text{OUT},\text{Th}} = 0.5 \times I_{\text{OUT},\text{target}}$ (typical)

9.3.2 Short-Circuit Detection Principle (TLC5917 Only)

The LED short-circuit detection compares the effective voltage level (V_{OUT}) with the shorted-load detection threshold voltages $V_{\rm OUT,TTh}$ and $V_{\rm OUT,RTh}$. If $V_{\rm OUT}$ is above the $V_{\rm OUT,TTh}$ threshold, the TLC5917 detects an shorted-load condition. If V_{OUT} is below the V_{OUT,RTh} threshold, no error is detected/error bit is reset. This error status can be read as an error status code in the Special Mode. For short-circuit error detection, a channel must be on.

Table 2. Shorted-Load Detection

Figure 10. Short-Circuit Detection Principle

9.3.3 Overtemperature Detection and Shutdown

TLC591x is equipped with a global overtemperature sensor and eight individual, channel-specific, overtemperature sensors.

- When the global sensor reaches the trip temperature, all output channels are shut down, and the error status is stored in the internal Error Status register of every channel. After shutdown, the channels automatically restart after cooling down, if the control signal (output latch) remains on. The stored error status is not reset after cooling down and can be read out as the error status code in the Special Mode.
- When one of the channel-specific sensors reaches trip temperature, only the affected output channel is shut down, and the error status is stored only in the internal Error Status register of the affected channel. After shutdown, the channel automatically restarts after cooling down, if the control signal (output latch) remains on. The stored error status is not reset after cooling down and can be read out as error status code in the Special Mode.

For channel-specific overtemperature error detection, a channel must be on.

The error status code is reset when TLC591x returns to Normal Mode.

Table 3. Overtemperature Detection(1)

(1) The global shutdown threshold temperature is approximately 170°C.

9.4 Device Functional Modes

The TLC591x provides two operation modes: Normal Mode and Special Mode. Normal mode is used for shifting LED data into and out of the driver. Special Mode includes two functions: Error Detection and Current Gain Control. The two operation modes include three phases: Normal Mode phase, Mode Switching transition phase, and Special Mode phase. The signal on the multiple function pin OE(ED2) is monitored to determine the mode. When a one-clock-wide pulse appears on OE(ED2), the device enters the Mode Switching phase. At this time, the voltage level on LE(ED1) determines which mode the TLC591x switches to.

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Device Functional Modes (continued)

Figure 11. Normal Mode

The signal sequence shown in [Figure](#page-16-1) 12 makes the TLC591x enter Current Adjust and Error Detection Mode.

Figure 12. Switching to Special Mode

In the Current Adjust Mode, sending the positive pulse of LE(ED1), the content of the shift register (a current adjust code) is written to the 8-bit configuration latch (see [Figure](#page-17-0) 13).

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Figure 13. Writing Configuration Code

When the TLC591x is in the Error Detection Mode, the signal sequence shown in [Figure](#page-17-1) 14 enables a system controller to read error status codes through SDO.

Figure 14. Reading Error Status Code

The signal sequence shown in [Figure](#page-17-2) 15 makes TLC591x resume the Normal Mode. Switching to Normal Mode resets all internal Error Status registers. \overline{OE} (ED2) always enables the output port, whether the TLC591x enters Current Adjust Mode or not.

Figure 15. Switching to Normal Mode

9.4.1 Operation Mode Switching

To switch between its two modes, TLC591x monitors the signal OE(ED2). When an one-clock-wide pulse of OE(ED2) appears, TLC591x enters the two-clock-period transition phase, the Mode Switching phase. After power on, the default operation mode is the Normal Mode (see [Figure](#page-18-0) 16).

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As shown in [Figure](#page-18-0) 16, once a one-clock-wide short pulse (101) of \overline{OE} (ED2) appears, TLC591x enters the Mode Switching phase. At the fourth rising edge of CLK, if LE(ED1) is sampled as voltage high, TLC591x switches to Special Mode; otherwise, it switches to Normal Mode. The signal LE(ED1) between the third and the fifth rising edges of CLK cannot latch any data. Its level is used only to determine into which mode to switch. However, the short pulse of OE(ED2) can still enable the output ports. During mode switching, the serial data can still be transferred through SDI and shifted out from SDO.

NOTE

- 1. The signal sequence for the mode switching may be used frequently to ensure that TLC591x is in the proper mode.
- 2. The 1 and 0 on the LE(ED1) signal are sampled at the rising edge of CLK. The X means its level does not affect the result of mode switching mechanism.
- 3. After power on, the default operation mode is Normal Mode.

9.4.1.1 Normal Mode Phase

Serial data is transferred into TLC591x through SDI, shifted in the Shift Register, and output via SDO. LE(ED1) can latch the serial data in the Shift Register to the Output Latch. $\overline{OE}(ED2)$ enables the output drivers to sink current. These functions differ only as described in Operation Mode Switching, in which case, a short pulse triggers TLC591x to switch the operation mode. However, as long as LE(ED1) is high in the Mode Switching phase, TLC591x remains in the Normal Mode, as if no mode switching occurred.

9.4.1.2 Special Mode Phase

In the Special Mode, as long as $\overline{OE}(ED2)$ is not low, the serial data is shifted to the Shift Register via SDI and shifted out via SDO, as in the Normal Mode. However, there are two differences between the Special Mode and the Normal Mode, as shown in the following sections.

9.4.2 Reading Error Status Code in Special Mode

When \overline{OE} (ED2) is pulled low while in Special Mode, error detection and load error status codes are loaded into the Shift Register, in addition to enabling output ports to sink current. [Figure](#page-19-0) 17 shows the timing sequence for error detection. The 0 and 1 signal levels are sampled at the rising edge of each CLK. At least three zeros must be sampled at the voltage low signal $\overline{OE}(\text{ED2})$. Immediately after the second zero is sampled, the data input source of the Shift Register changes to the 8-bit parallel Error Status Code register, instead of from the serial data on SDI. Normally, the error status codes are generated at least 2 μs after the falling edge of OE(ED2). The occurrence of the third or later zero saves the detected error status codes into the Shift Register. Therefore, when OE(ED2) is low, the serial data cannot be shifted into TLC591x through SDI. When OE(ED2) is pulled high, the data input source of the Shift Register is changed back to SDI. At the same time, the output ports are disabled and the error detection is completed. Then, the error status codes saved in the Shift Register can be shifted out via SDO bit by bit along with CLK, as well as the new serial data can be shifted into TLC591x through SDI.

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FXAS

While in Special Mode, the TLC591x cannot simultaneously transfer serial data and detect LED load error status.

Figure 17. Reading Error Status Code

9.4.3 Writing Configuration Code in Special Mode

When in Special Mode, the active high signal LE(ED1) latches the serial data in the Shift Register to the Configuration Latch, instead of the Output Latch. The latched serial data is used as the Configuration Code.

The code is stored until power off or the Configuration Latch is rewritten. As shown in [Figure](#page-19-1) 18, the timing for writing the Configuration Code is the same as the timing in the Normal Mode to latching output channel data. Both the Configuration Code and Error Status Code are transferred in the common 8-bit Shift Register. Users must pay attention to the sequence of error detection and current adjustment to avoid the Configuration Code being overwritten by Error Status Code.

Figure 18. Writing Configuration Code

10 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

10.1 Application Information

10.1.1 Constant Current

In LED display applications, TLC591x provides nearly no current variations from channel to channel and from IC to IC. While 5 mA $\leq I_{\text{OUT}} \leq 100$ mA, the maximum current skew between channels is less than $\pm 3\%$ and between ICs is less than ±6%.

10.1.2 Adjusting Output Current

TLC591x scales up the reference current, I_{ref} , set by the external resistor R_{ext} to sink a current, I_{out} , at each output port. Users can follow the below formulas to calculate the target output current $I_{\text{OUT.target}}$ in the saturation region. In the equations,

 R_{ext} is the resistance of the external resistor connected between the R-EXT terminal and ground and V_{R-EXT} is the voltage of R-EXT, which is controlled by the programmable voltage gain (VG). VG is defined by the Configuration Code.

The Current Multiplier (CM) determines that the ratio $I_{OUT,target}/I_{ref}$ is 15 or 5. After power on, the default value of VG is 127/128 = 0.992, and the default value of CM is 1, so that the ratio $I_{OUT,target}/I_{ref}$ = 15. Based on the default VG and CM:

$$
V_{R\text{-}EXT} = 1.26 \text{ V} \times 127/128 = 1.25 \text{ V}
$$
\n
$$
I_{\text{OUT.target}} = (1.25 \text{ V/R}_{\text{ext}}) \times 15 \tag{4}
$$

Therefore, the default current is approximately 52 mA at 360 Ω and 26 mA at 720 Ω . The default relationship after power on between $I_{\text{OUT:target}}$ and R_{ext} is shown in [Figure](#page-20-2) 19.

Figure 19. Default Relationship Curve Between IOUT,target and Rext After Power Up

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Application Information (continued)

10.1.3 Cascading Implementation of TLC591x Device

Figure 20. Cascading Implementation of TLC591x Device

Application Information (continued)

10.1.4 8-Bit Configuration Code and Current Gain

Bit definition of the Configuration Code in the Configuration Latch is shown in [Table](#page-22-0) 5.

		. .						
Meaning	CМ	HC	CC ₀	\sim ◡◡╷	CC ₂	\sim ບບວ	CC4	CC ₅
Default								

Table 5. Bit Definition of 8-Bit Configuration Code

Bit 7 is first sent into TLC591x through SDI. Bits 1 to 7 {HC, CC[0:5]} determine the voltage gain (VG) that affects the voltage at R-EXT and indirectly affects the reference current, I_{ref}, flowing through the external resistor at R-EXT. Bit 0 is the Current Multiplier (CM) that determines the ratio $I_{OUT,target}/I_{ref}$. Each combination of VG and CM gives a specific Current Gain (CG).

• VG: the relationship between {HC,CC[0:5]} and the voltage gain is calculated as shown in [Equation](#page-22-1) 6 and [Equation](#page-22-2) 7:

$$
VG = (1 + HC) \times (1 + D/64) / 4
$$
 (6)

D = CC0 \times 2⁵ + CC1 \times 2⁴ + CC2 \times 2³ + CC3 \times 2² + CC4 \times 2¹ + CC5 \times 2⁰

(7)

Where HC is 1 or 0, and D is the binary value of CC[0:5]. So, the VG could be regarded as a floating-point number with 1-bit exponent HC and 6-bit mantissa CC[0:5]. {HC,CC[0:5]} divides the programmable voltage gain VG into 128 steps and two sub-bands:

Low voltage sub-band (HC = 0): $VG = 1/4 \sim 127/256$, linearly divided into 64 steps

High voltage sub-band (HC = 1): $VG = 1/2 \sim 127/128$, linearly divided into 64 steps CM: In addition to determining the ratio $I_{\text{OUT target}}/I_{\text{ref}}$, CM limits the output current range.

High Current Multiplier (CM = 1): $I_{\text{OUT,target}}/I_{\text{ref}} = 15$, suitable for output current range $I_{\text{OUT}} = 10$ mA to 120 mA. Low Current Multiplier (CM = 0): $I_{OUT,target}/I_{ref} = 5$, suitable for output current range $I_{OUT} = 3$ mA to 40 mA

• CG: The total Current Gain is defined as the following.

 $V_{R-FXT} = 1.26 \text{ V} \times \text{VG}$ (8)

 $I_{ref} = V_{R-EXT}/R_{ext}$, if the external resistor, R_{ext} , is connected to ground. (9)

 $I_{\text{OUT,target}} = I_{\text{ref}} \times 15 \times 3^{\text{CM} - 1} = 1.26 \text{ V/R}_{\text{ext}} \times \text{VG} \times 15 \times 3^{\text{CM} - 1} = (1.26 \text{ V/R}_{\text{ext}} \times 15) \times \text{CG}$ (10) $CG = VG \times 3^{CM-1}$ (11)

Therefore, CG = (1/12) to (127/128), and it is divided into 256 steps. If CG = 127/128 = 0.992, the $I_{\text{OUT target}}$ R_{ext} .

Examples

- Configuration Code {CM, HC, $CC[0:5]$ } = {1,1,111111} $VG = 127/128 = 0.992$ and $CG = VG \times 3^0 = VG = 0.992$
- Configuration Code = $\{1,1,000000\}$ $VG = (1 + 1) \times (1 + 0/64)/4 = 1/2 = 0.5$, and $CG = 0.5$
- **Configuration Code =** $\{0.0,000000\}$ $VG = (1 + 0) \times (1 + 0/64)/4 = 1/4$, and $CG = (1/4) \times 3^{-1} = 1/12$

After power on, the default value of the Configuration Code {CM, HC, CC[0:5]} is {1,1,111111}. Therefore, $VG = CG = 0.992$. The relationship between the Configuration Code and the Current Gain is shown in [Figure](#page-23-2) 21.

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Figure 21. Current Gain vs Configuration Code

10.2 Typical Application

[Figure](#page-23-1) 22 shows implementation of a single TLC591x device. [Figure](#page-21-0) 20 shows a cascaded driver implementation.

Figure 22. Single Implementation of TLC591x Device

Typical Application (continued)

10.2.1 Design Requirements

For this design example, use the parameters listed in [Table](#page-24-0) 6. The purpose of this design procedure is to calculate the power dissipation in the device and the operating junction temperature.

10.2.2 Detailed Design Procedure

 $T_J = T_A + R_{\theta JA} \times P_{D\, \text{TOT}}$

where

- T_{J} is the junction temperature.
- T_A is the ambient temperature.
- R_{theta} is the junction-to-ambient thermal resistance.
- $P_{\text{D} \text{ TOT}}$ is the total power dissipation in the IC. (12) $P_{\text{D} \text{ TOT}} = P_{\text{D} \text{ CS}} + I_{\text{DD}} \times V_{\text{DD}}$ where • P_{D} csis the power dissipation in the LED current sinks. \bullet I_{DD} is the IC supply current. • V_{DD} is the IC supply voltage. (13) $P_{D_{CSS}} = I_0 \times V_0 \times n_{CH}$

where

- V_{Ω} is the voltage at the output pin.
- n_{CH} is the number of LED strings. (14)

$V_O = V_{LED} - (n_{LED} \times V_F)$

where

- \bullet V_{LED} is the voltage applied to the LED string.
- n_{LED} is the number of LEDs in the string.
- V_F is the forward voltage of each LED. (15) (15)

 V_O must not be too high as this causes excess power dissipation inside the current sink. However, V_O also must not be too low as this does not allow the full LED current ([Figure](#page-8-2) 4). With $V_{LED} = 12$ V:

Using P_{D_TOT} , calculate:

$$
T_J = T_A + R_{\theta JA} \times P_{D_TOT} = 115^{\circ}\text{C} + 87.4^{\circ}\text{C/W} \times 0.29 \text{ W} = 140^{\circ}\text{C}
$$
 (19)

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This design example demonstrates how to calculate power dissipation in the IC and ensure that the junction temperature is kept below 150°C.

NOTE

This design example assumes that all channels have the same electrical parameters $(n_{LED}, I_0, V_F, V_{LED})$. If the parameters are unique for each channel, then the power dissipation must be calculated for each current sink separately. Then, each result must be added together to calculate the total power dissipation in the current sinks.

10.2.3 Application Curve

Figure 23. Output Current vs Output Voltage

11 Power Supply Recommendations

The device is designed to operate from a VDD supply between 3 V and 5.5 V. The LED supply voltage is determined by the number of LEDs in each string and the forward voltage of the LEDs.

12 Layout

12.1 Layout Guidelines

The traces that carry current from the LED cathodes to the OUTx pins must be wide enough to support the default current (up to 120 mA).

The SDI, CLK, LE (ED1), OE (ED2), and SDO pins are to be connected to the microcontroller. There are several ways to achieve this, including the following methods:

- Traces may be routed underneath the package on the top layer.
- The signal may travel through a via to another layer.

12.2 Layout Example

VIA to GND

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Layout Example (continued)

13 Device and Documentation Support

13.1 Related Links

The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

Table 7. Related Links

13.2 Trademarks

All trademarks are the property of their respective owners.

13.3 Electrostatic Discharge Caution

These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

13.4 Glossary

[SLYZ022](http://www.ti.com/lit/pdf/SLYZ022) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

14 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check<http://www.ti.com/productcontent>for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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OTHER QUALIFIED VERSIONS OF TLC5916, TLC5917 :

PACKAGE OPTION ADDENDUM

• Automotive: [TLC5916-Q1](http://focus.ti.com/docs/prod/folders/print/tlc5916-q1.html), [TLC5917-Q1](http://focus.ti.com/docs/prod/folders/print/tlc5917-q1.html)

NOTE: Qualified Version Definitions:

• Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

PACKAGE MATERIALS INFORMATION

Texas
Instruments

TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

TEXAS
SINSTRUMENTS

www.ti.com 1-Oct-2014

PACKAGE MATERIALS INFORMATION

*All dimensions are nominal

 $D (R-PDSO-G16)$

PLASTIC SMALL OUTLINE

NOTES: A. All linear dimensions are in inches (millimeters).

- B. This drawing is subject to change without notice.
- 6 Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AC.

4211283-4/E 08/12

$D (R - PDSO - G16)$ PLASTIC SMALL OUTLINE Stencil Openings
(Note D) Example Board Layout (Note C) $-16x0,55$ $-14x1,27$ $-14x1,27$ 16x1,50 $5,40$ 5.40 Example Non Soldermask Defined Pad Example Pad Geometry (See Note C) $-0,60$ 1.55 Example Solder Mask Opening (See Note E) $-0,07$

NOTES: A. All linear dimensions are in millimeters.

All Around

- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations. E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

PW (R-PDSO-G16)

PLASTIC SMALL OUTLINE

This drawing is subject to change without notice. **B.**

 $\hat{\mathbb{C}}$ Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.

 $\hat{\mathbb{D}}$ Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.

E. Falls within JEDEC MO-153

PW (R-PDSO-G16)

PLASTIC SMALL OUTLINE

- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

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